

# ECEN689: Special Topics in High-Speed Links Circuits and Systems Spring 2010

---

## Lecture 7: Channel Time-Domain Response



Sam Palermo

Analog & Mixed-Signal Center

Texas A&M University

# Announcements

---

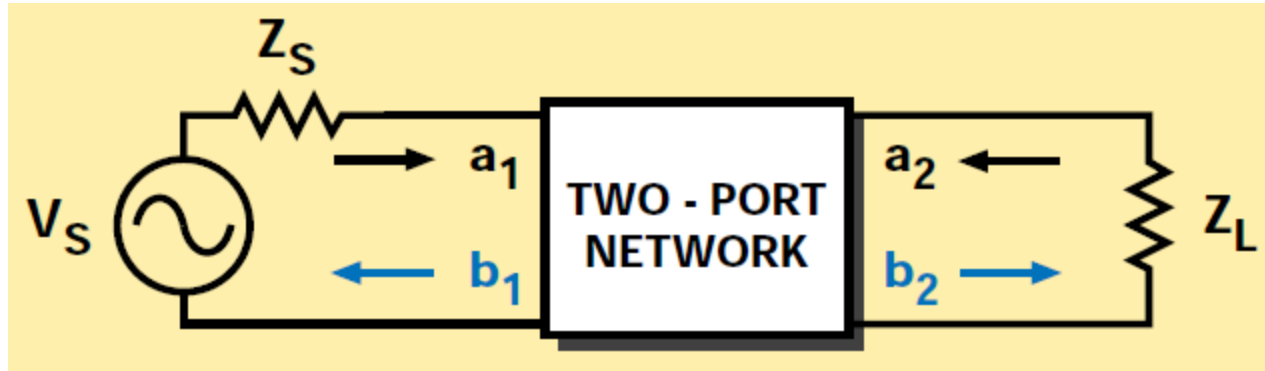
- HW2 due 2/5
- **No class next week**
- Reading
  - Will post some material on TDR and network analyzers (S-parameters)
  - Link signaling papers

# Agenda

---

- S-parameters revisited
- Impulse response generation
- Eye diagrams
- Inter-symbol interference (ISI)

# Formal S-Parameter Definitions



[Agilent]

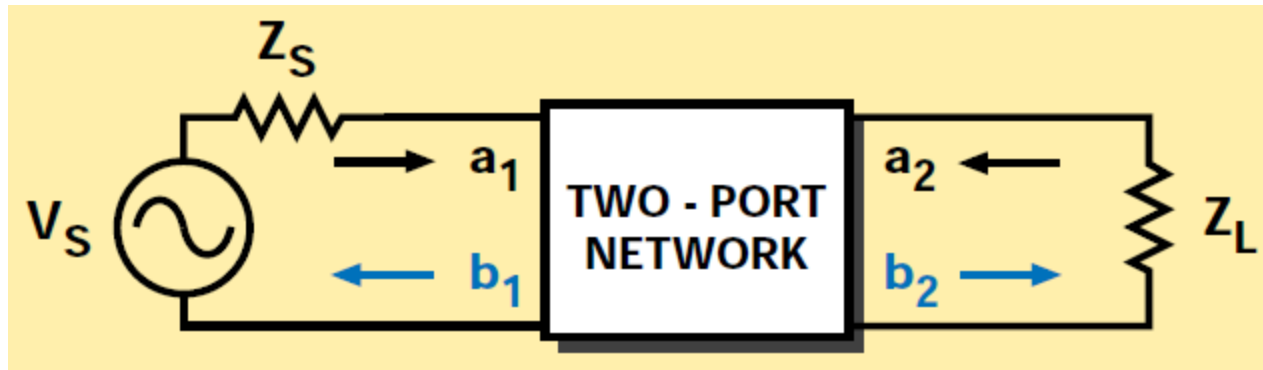
$$s_{11} = \left. \frac{b_1}{a_1} \right|_{a_2=0} = \text{Input reflection coefficient with the output port terminated by a matched load } (Z_L = Z_0 \text{ sets } a_2=0)$$

$$s_{22} = \left. \frac{b_2}{a_2} \right|_{a_1=0} = \text{Output reflection coefficient with the input terminated by a matched load } (Z_S = Z_0 \text{ sets } V_S=0)$$

$$s_{21} = \left. \frac{b_2}{a_1} \right|_{a_2=0} = \text{Forward transmission (insertion) gain with the output port terminated in a matched load.}$$

$$s_{12} = \left. \frac{b_1}{a_2} \right|_{a_1=0} = \text{Reverse transmission (insertion) gain with the input port terminated in a matched load.}$$

# S-Parameters with Arbitrary Termination



[Agilent]

Input reflection coefficient with arbitrary  $Z_L$

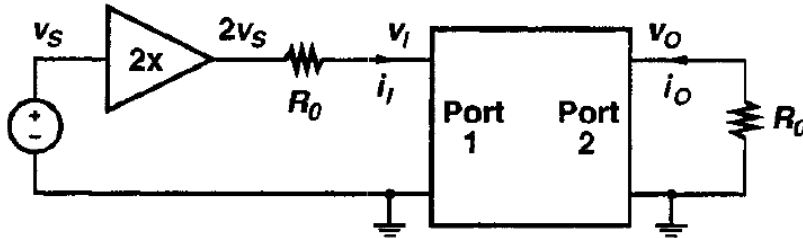
$$s'_{11} = s_{11} + \frac{s_{12} s_{21} \Gamma_L}{1 - s_{22} \Gamma_L} \quad (\mathbf{a_2 \neq 0})$$

Output reflection coefficient with arbitrary  $Z_S$

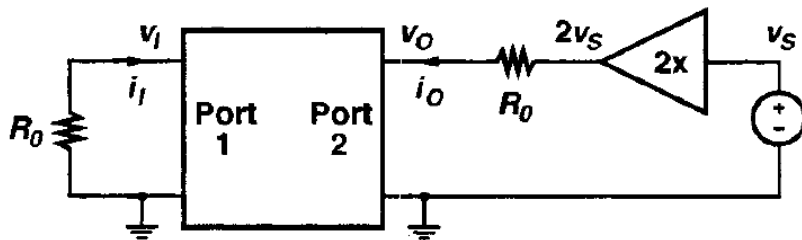
$$s'_{22} = s_{22} + \frac{s_{12} s_{21} \Gamma_S}{1 - s_{11} \Gamma_S} \quad (\mathbf{a_1 \neq 0})$$

- I believe this is what the network analyzer reports for  $s_{11}$  and  $s_{22}$  when a matching network is not used

# S-Parameter Test Circuits & Meaning



[Sackinger]



$$S_{11}(s) = \frac{V_{i,reflected}}{V_{i,incident}} = \frac{V_i - R_0 I_i}{V_i + R_0 I_i} = \frac{V_i - V_s}{V_s}$$

$$S_{21}(s) = \frac{V_{o,transmitted}}{V_{i,incident}} = \frac{V_o - R_0 I_o}{V_i + R_0 I_i} = \frac{V_o}{V_s}$$

$$S_{22}(s) = \frac{V_{o,reflected}}{V_{o,incident}} = \frac{V_o - R_0 I_o}{V_o + R_0 I_o} = \frac{V_o - V_s}{V_s}$$

$$S_{12}(s) = \frac{V_{i,transmitted}}{V_{o,incident}} = \frac{V_i - R_0 I_i}{V_o + R_0 I_o} = \frac{V_i}{V_s}$$

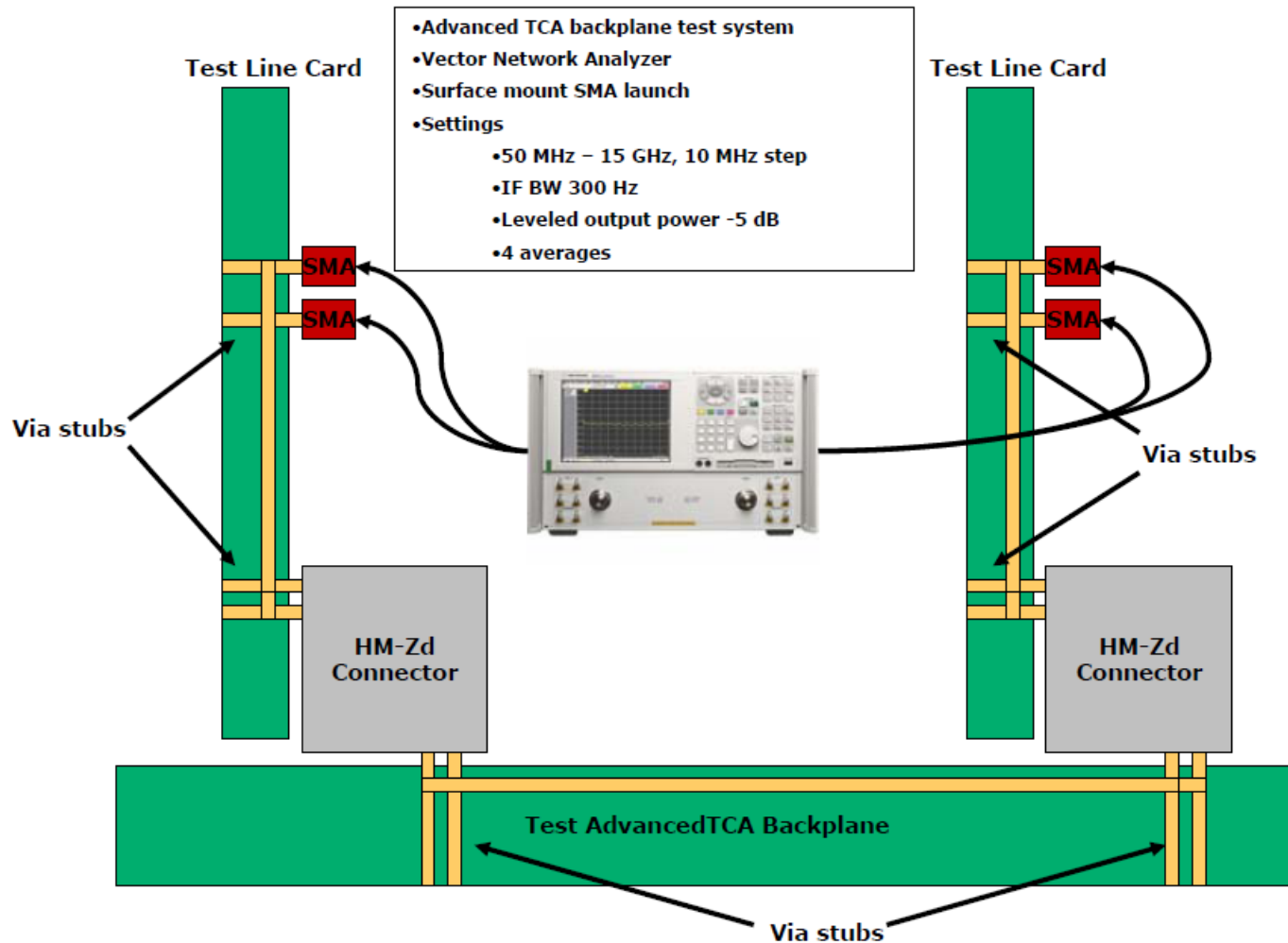
$$S_{21}(s) = [1 + S_{11}(s)]A(s)$$

where A(s) is loaded voltage gain

- $S_{11}$  = Input reflection coefficient
  - $1/S_{11}$  = Input return loss
- $S_{21}$  = Forward transmission coefficient
  - Gain w/ input matching dependency
- $S_{22}$  = Output reflection coefficient
  - $1/S_{22}$  = Output return loss
- $S_{12}$  = Reverse transmission coefficient (isolation)

**If a1 and a2 are not equal to zero for the appropriate measurements, these are "formally"  $S'_{11}$ ,  $S'_{21}$ ,  $S'_{22}$ ,  $S'_{12}$**

# S-Parameter Channel Example



[Peters, IEEE Backplane Ethernet Task Force]

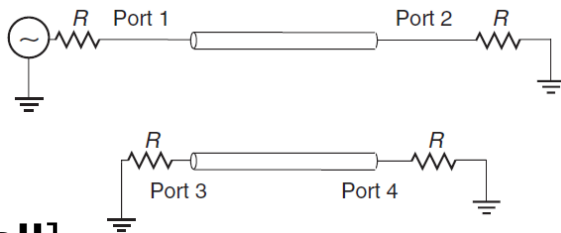
# S-Parameter Channel Example (4-port differential)

```
! peters_01_0605 rzu channel thru response
#
! HZ S RI R 50
!
! FREQ S11 S12 S13 S14
! S21 S22 S23 S24
! S31 S32 S33 S34
! S41 S42 S43 S44
!
! REAL IMAG REAL IMAG REAL IMAG REAL IMAG REAL IMAG
5.00000000e+007 6.279266901548e-002 -5.256007502766e-002 -1.995363973143e-001 -9.018006169275e-001 7.405252014369e-002 -1.653914717779e-002 4.694410796534e-004 2.855671737566e-003
-1.993592781969e-001 -9.017752677900e-001 6.847049395661e-002 -3.537762509466e-002 6.592975593456e-004 2.600733690373e-003 7.478976460177e-002 -1.488182269791e-002
7.438370524663e-002 -1.650568516548e-002 6.663957537997e-004 2.723661634513e-003 5.641343731365e-002 -5.693035832892e-002 -2.070369894915e-001 -8.986367167361e-001
3.380698172980e-004 2.715033111885e-003 7.497765935351e-002 -1.488546535615e-002 -2.063544808970e-001 -9.002700655374e-001 6.856095801756e-002 -3.019606086420e-002
6.00000000e+007 4.829977376755e-002 -6.288238652440e-002 -4.923832497425e-001 -7.721510464035e-001 6.298956599590e-002 -3.938489680891e-002 1.125377257145e-003 1.921732299021e-003
-4.925547500023e-001 -7.726263821707e-001 6.163450406360e-002 -4.486265928179e-002 1.299644022342e-003 1.492436402394e-003 6.462146347807e-002 -3.736630924981e-002
6.308085276969e-002 -3.947655302643e-002 1.386741613180e-003 1.653454474207e-003 4.393874455850e-002 -6.448913049207e-002 -4.992743919180e-001 -7.660808533046e-001
1.280875740087e-003 1.936760526874e-003 6.482369657086e-002 -3.743006383763e-002 -4.995203164654e-001 -7.674804458241e-001 6.284893613667e-002 -4.132139739274e-002
```

Data from 50MHz to 15GHz in  
10MHz steps



```
1.49900000e+010 -1.884123481138e-001 3.522933794755e-001 9.493645552321e-004 2.735890006358e-004 2.939002692375e-002 -8.676465491258e-003 -2.207496924854e-004 1.236065259912e-004
9.463443060684e-004 3.105615146344e-004 -1.742347383703e-001 4.813685271232e-002 -6.152705437030e-004 1.614752661571e-003 6.774475978813e-002 9.617239585695e-003
2.953403898205e-002 -8.707827389646e-003 -6.226849675423e-004 1.637610280621e-003 -1.595766021694e-001 3.757605914955e-001 -1.809501624148e-004 -7.061855554470e-004
-2.613575703191e-004 1.368108929760e-004 6.788329666403e-002 9.551687705500e-003 -2.146293806886e-004 -7.363580847286e-004 -1.199804891859e-001 7.697336952293e-002
1.50000000e+010 -1.883176013184e-001 3.545614742110e-001 9.524680768441e-004 -5.404222971799e-005 2.935126165241e-002 -1.235086132268e-002 -1.616280086909e-004 2.3473368458649e-004
1.039250921080e-003 -6.032017103742e-005 -1.649137634331e-001 4.966164587830e-002 -6.748937194262e-005 1.689652681670e-003 6.725041473699e-002 1.961009613152e-003
2.959693594806e-002 -1.251203706381e-002 -2.927441863297e-005 1.747754847916e-003 -1.531702433245e-001 3.773014940454e-001 -3.769459376261e-004 -5.671620228005e-004
-2.089293612250e-004 2.303682313561e-004 6.740524959192e-002 1.672663579641e-003 -4.385850073691e-004 -5.810569604703e-004 -1.121319455376e-001 7.458173831411e-002
```



[Hall]

$$\begin{bmatrix} b_1 \\ b_2 \\ b_3 \\ b_4 \end{bmatrix} = \begin{bmatrix} S_{11} & S_{12} & S_{13} & S_{14} \\ S_{21} & S_{22} & S_{23} & S_{24} \\ S_{31} & S_{32} & S_{33} & S_{34} \\ S_{41} & S_{42} & S_{43} & S_{44} \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \\ a_3 \\ a_4 \end{bmatrix} = \begin{bmatrix} S_{11} & S_{12} & S_{13} & S_{14} \\ S_{21} & S_{22} & S_{23} & S_{24} \\ S_{31} & S_{32} & S_{33} & S_{34} \\ S_{41} & S_{42} & S_{43} & S_{44} \end{bmatrix} \begin{bmatrix} v \\ 0 \\ -v \\ 0 \end{bmatrix}$$

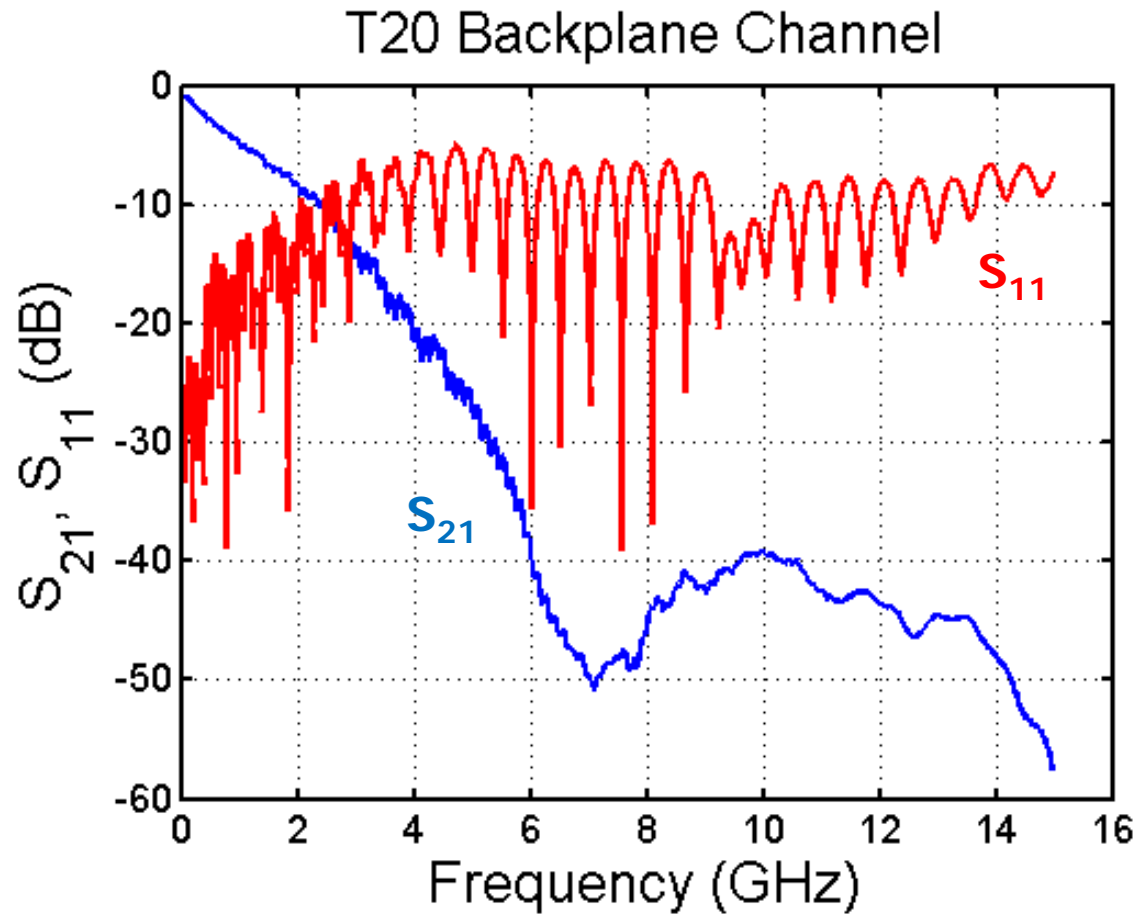
$$S_{dd11} = \left. \frac{b_{d1}}{a_{d1}} \right|_{a_2=a_4=0} = \frac{1}{2}(S_{11} + S_{33} - S_{13} - S_{31})$$

$$S_{dd21} = \left. \frac{b_{d2}}{a_{d1}} \right|_{a_2=a_4=0} = \frac{1}{2}(S_{21} + S_{43} - S_{23} - S_{41})$$



# S-Parameter Channel Example

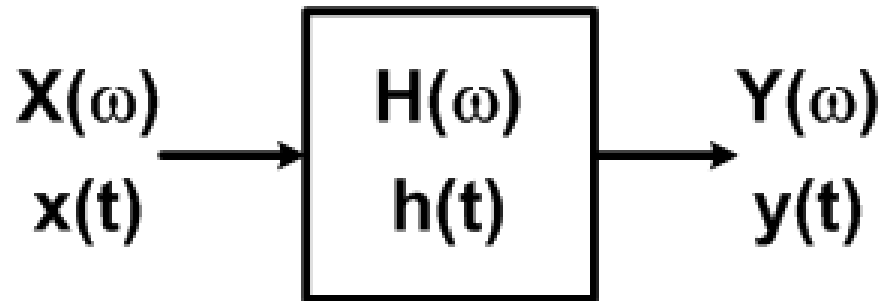
---



# Impulse Response

---

- Channel impulse responses are used in
  - Time domain simulations
  - Link analysis tools



$$Y(\omega) = H(\omega)X(\omega)$$

$$y(t) = h(t) * x(t) = \int_{-\infty}^{\infty} h(t - \tau)x(\tau) d\tau$$

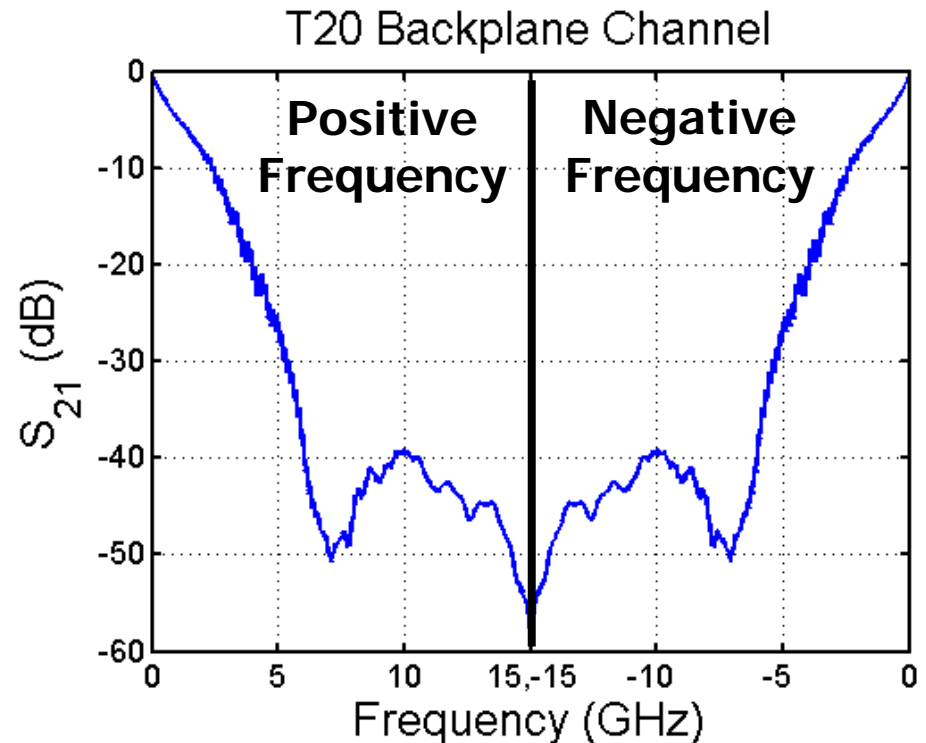
$$h(t) = F^{-1}\{H(\omega)\}$$

# Generating an Impulse Response from S-Parameters

- Perform the inverse Fourier transform on the s-parameter of interest
- Step 1: For ifft, produce negative frequency values and append to s-parameter data in the following manner

$$S(-f) = S(f)^*$$

$$h(t) = F^{-1} \{S(\omega)\}$$



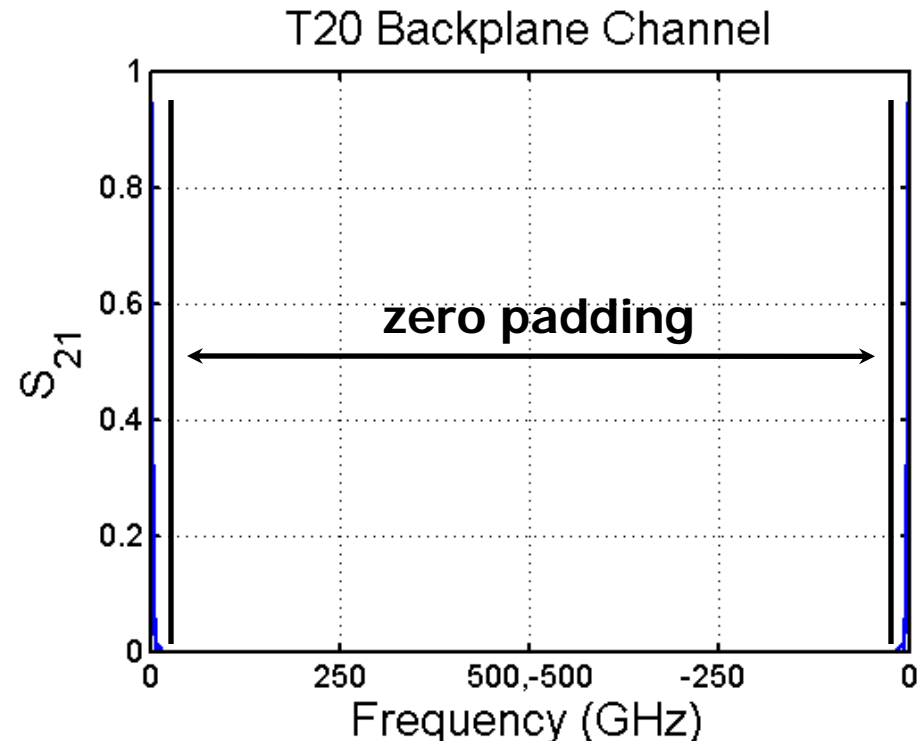
# Increasing Impulse Response Resolution

- Could perform ifft now, but will get an impulse response with time resolution of

$$\frac{1}{2f_{\max}} = \frac{1}{2(15\text{GHz})} = 33.3\text{ps}$$

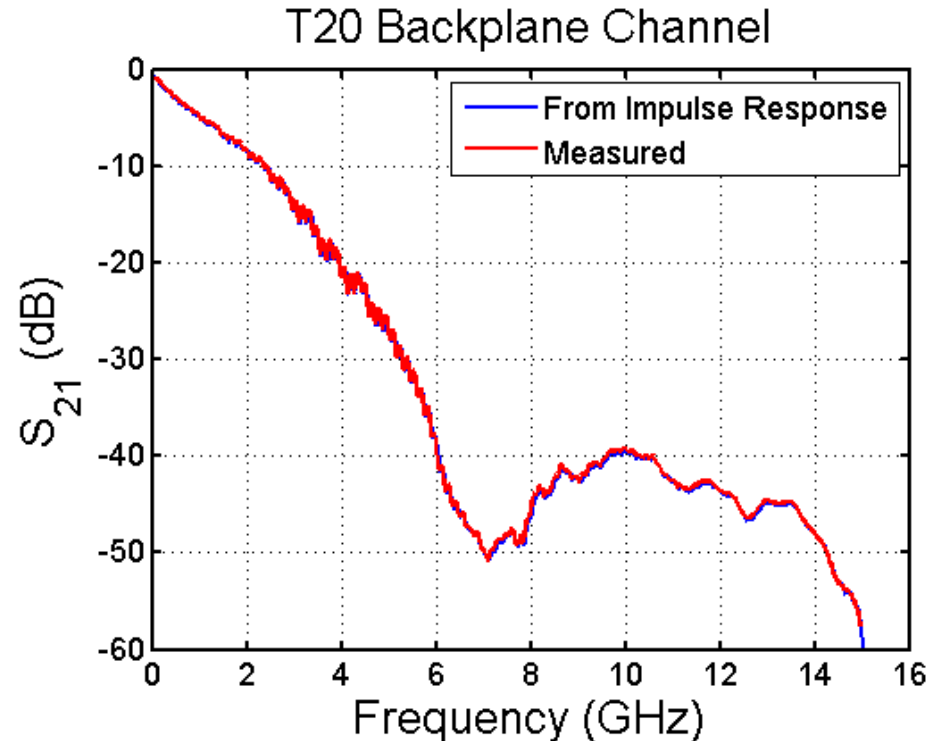
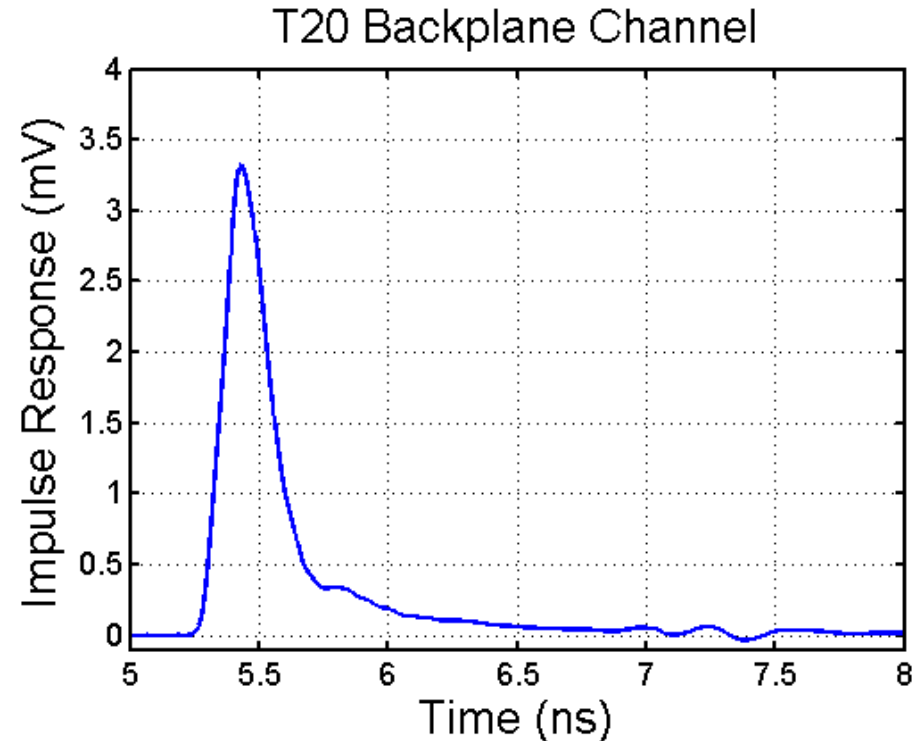
- To improve impulse response resolution expand frequency axis and “zero pad”

For 1ps resolution:  
zero pad to +/-500GHz



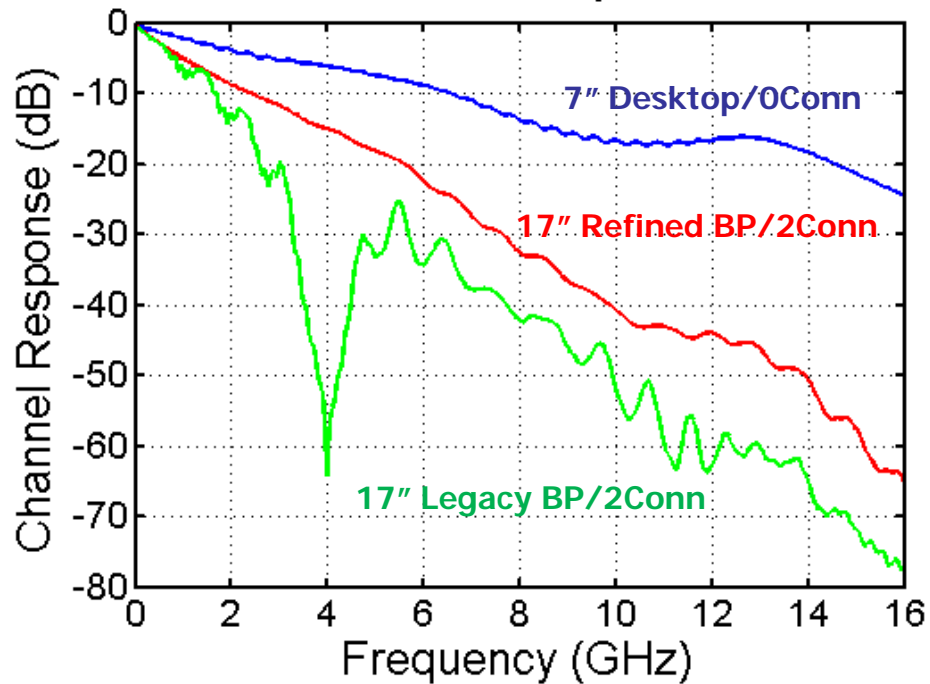
# Channel Impulse Response

- Now perform ifft to produce impulse response
- Can sanity check by doing an fft on impulse response and comparing to measured data

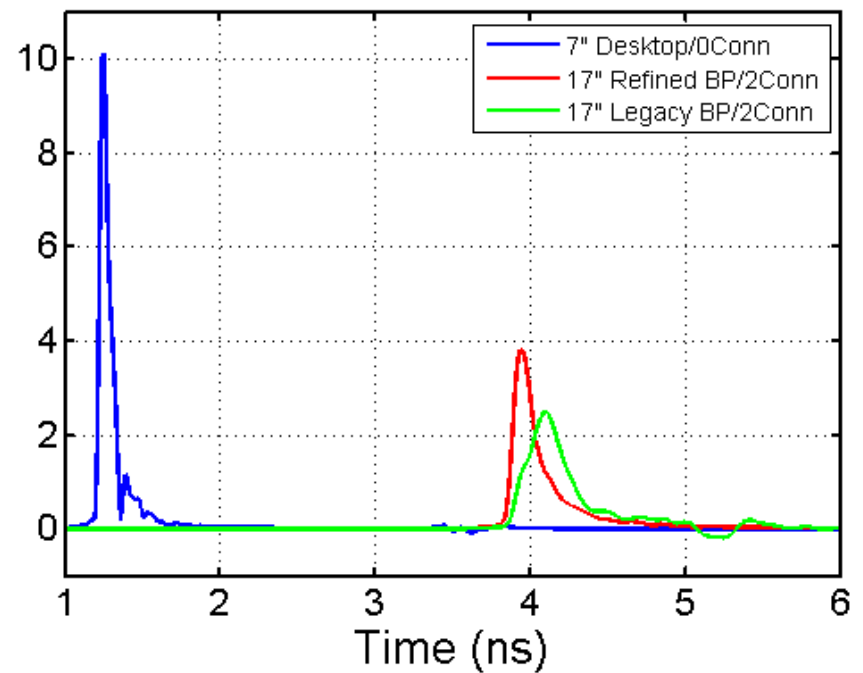


# Impulse Response of Different Channels

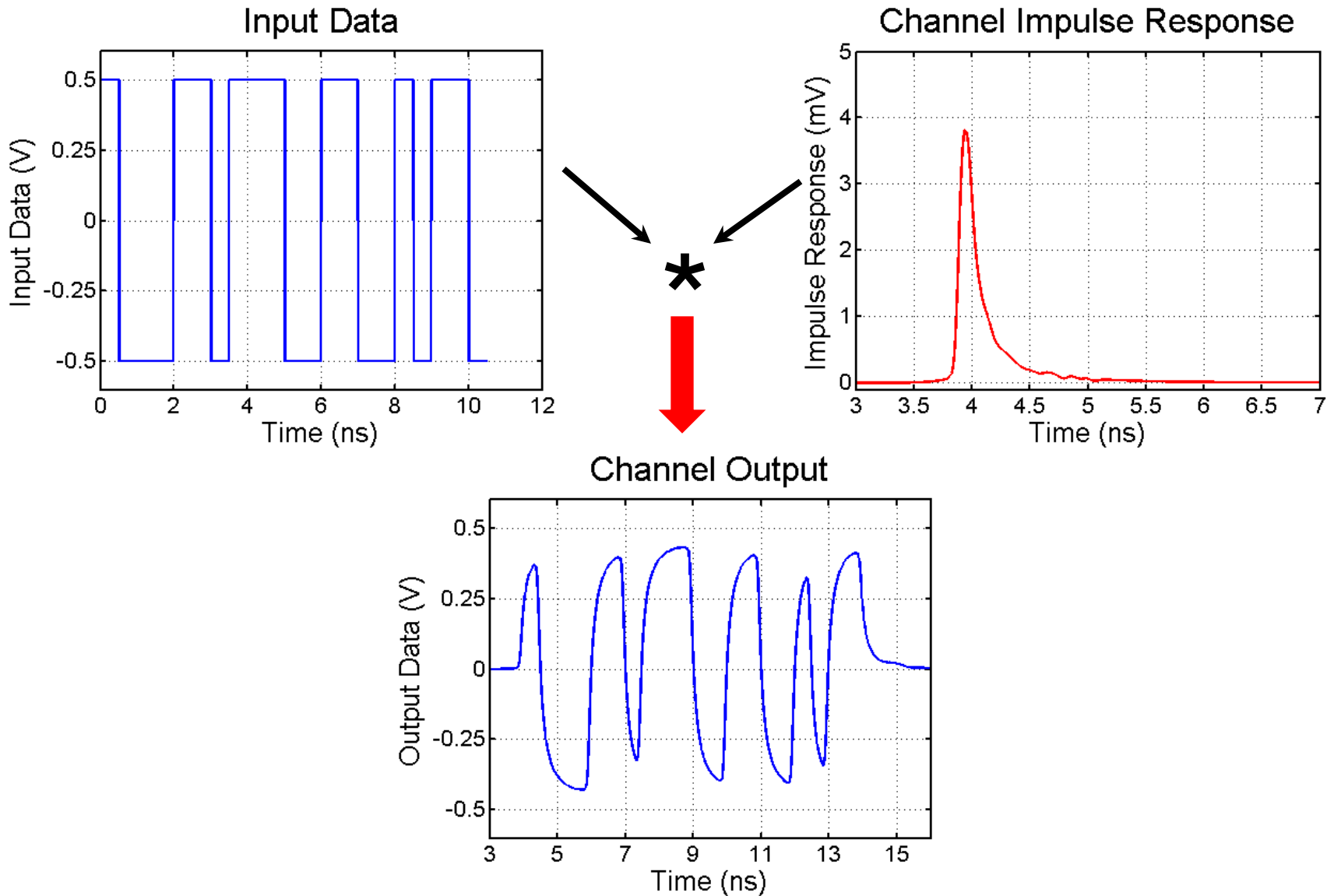
## Channel Responses



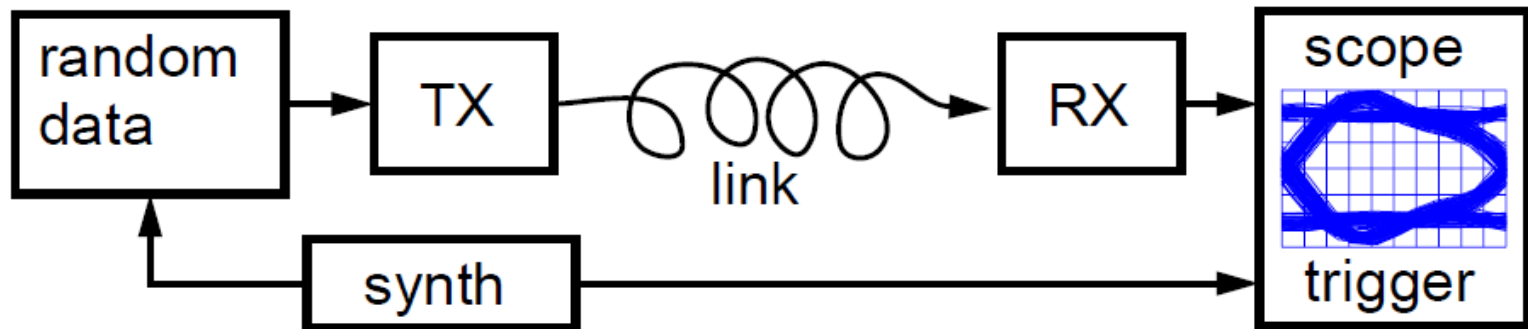
## Channel Impulse Responses



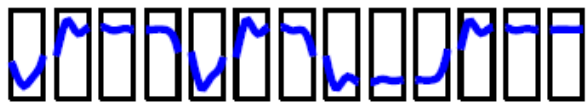
# Channel Transient Response



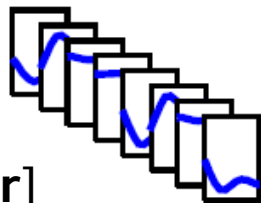
# Eye Diagrams



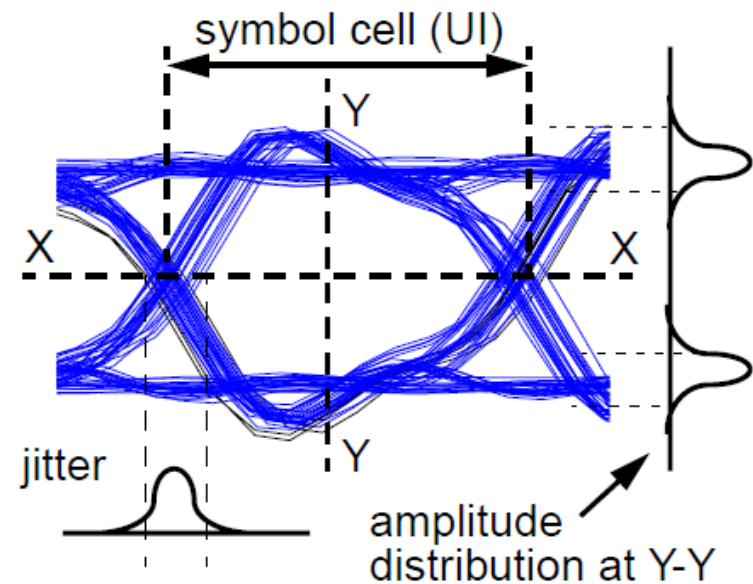
Use a precise clock to chop the data into equal periods



overlay each period onto one plot



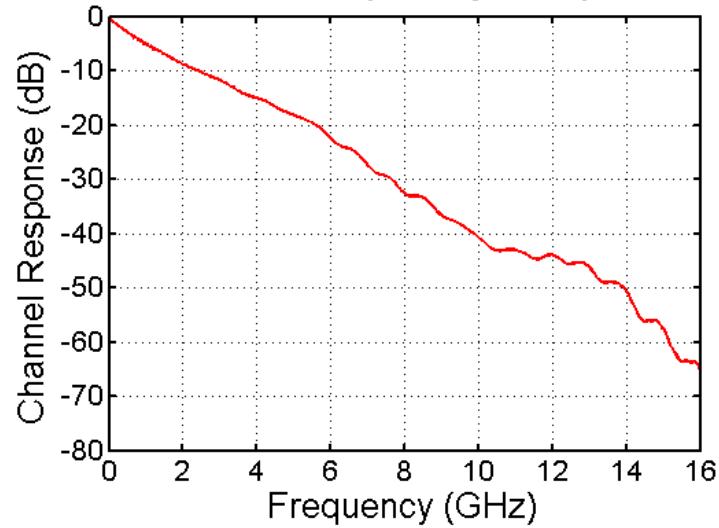
[Walker]



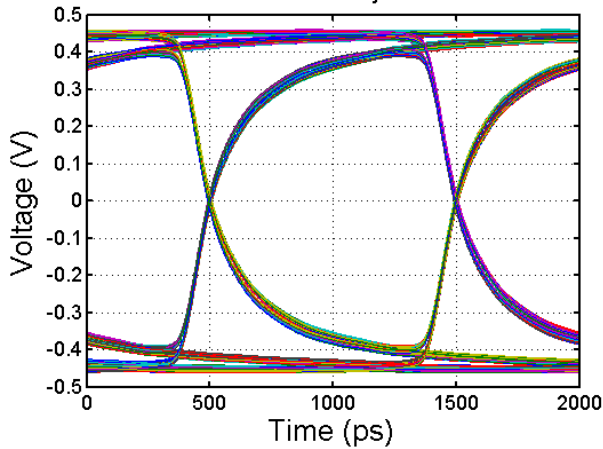


# Eye Diagrams vs Data Rate

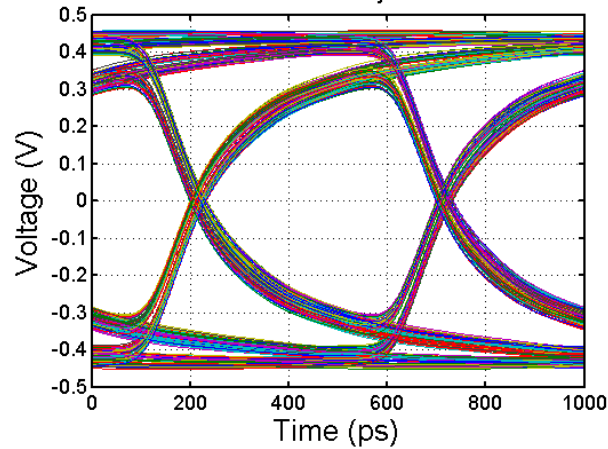
## Channel Frequency Response



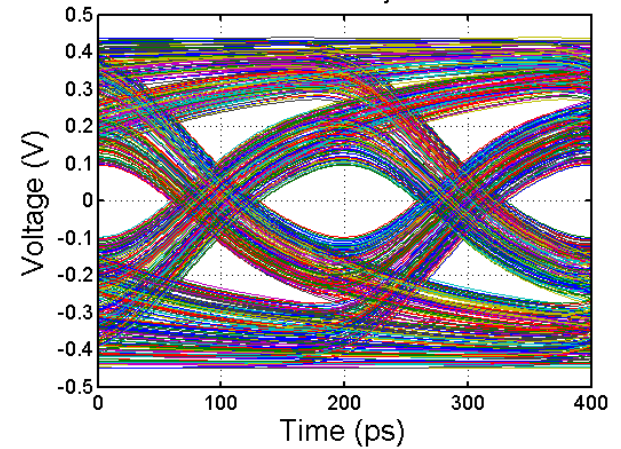
## 1Gb/s Eye



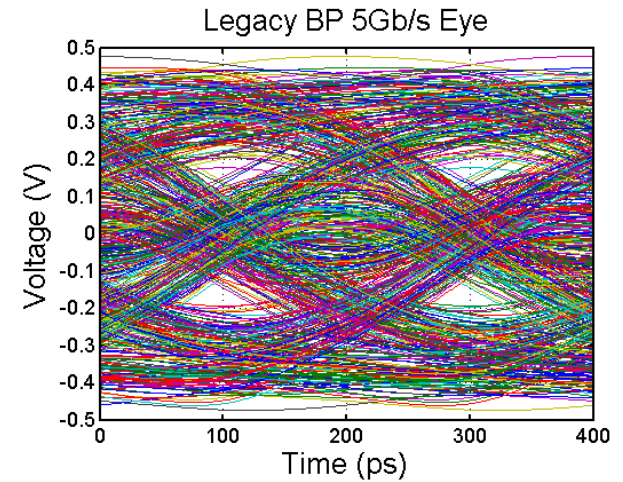
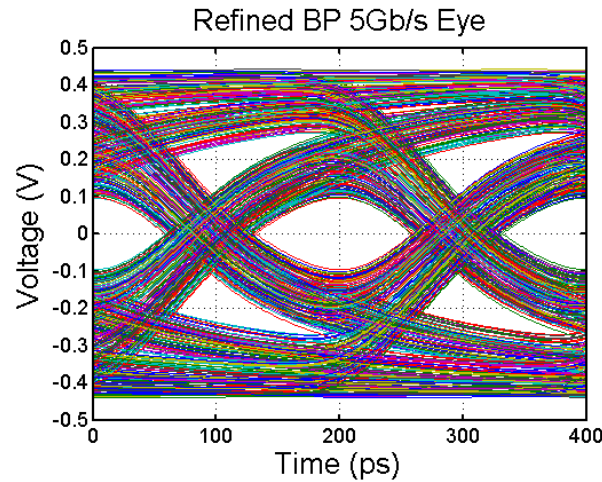
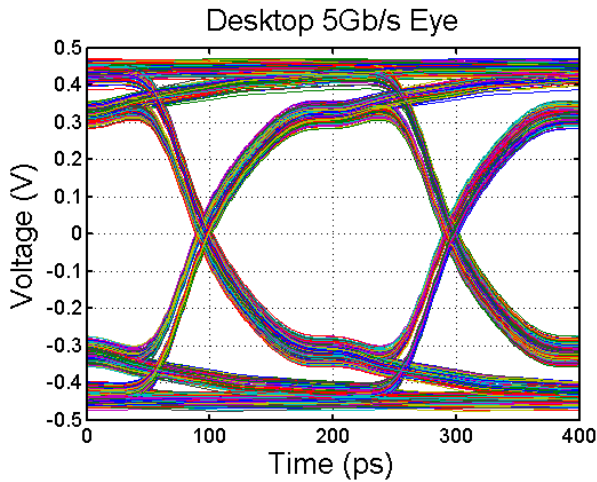
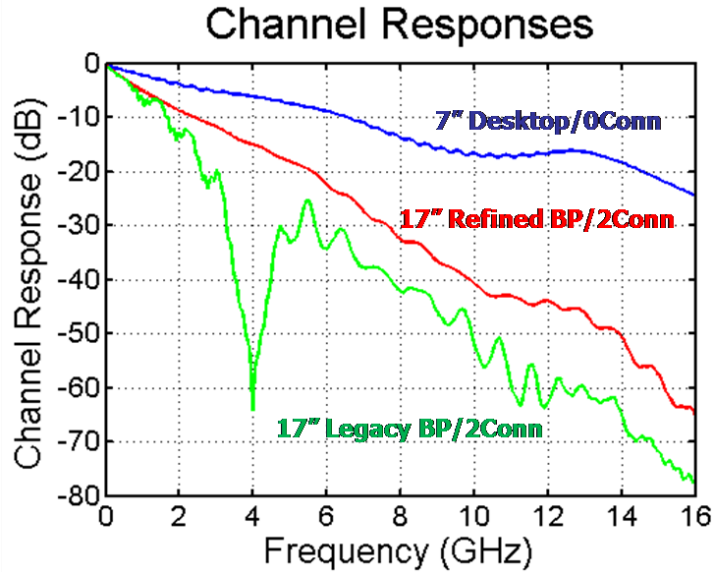
## 2Gb/s Eye



## 5Gb/s Eye



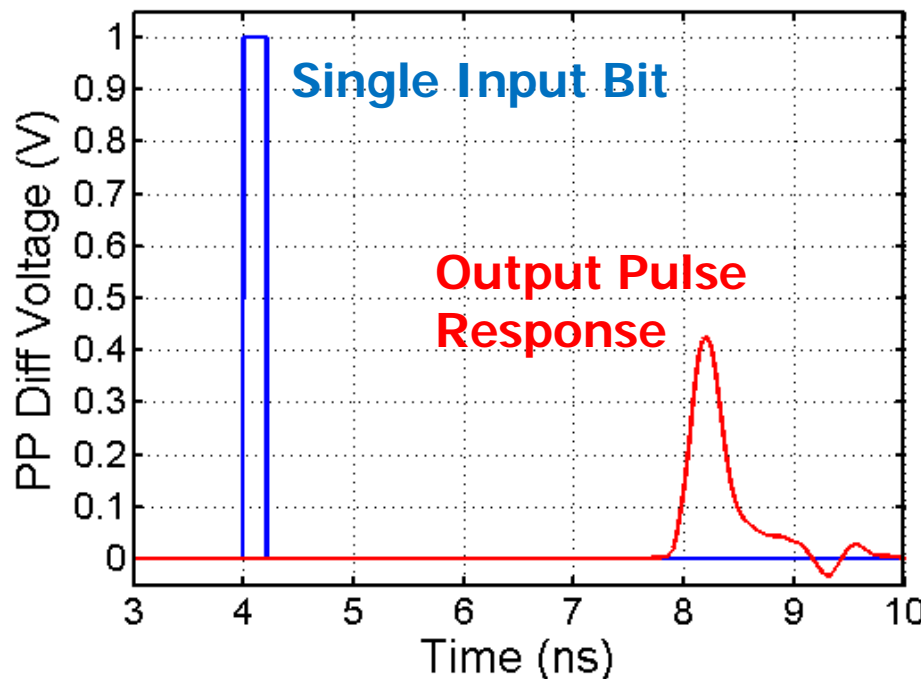
# Eye Diagrams vs Channel



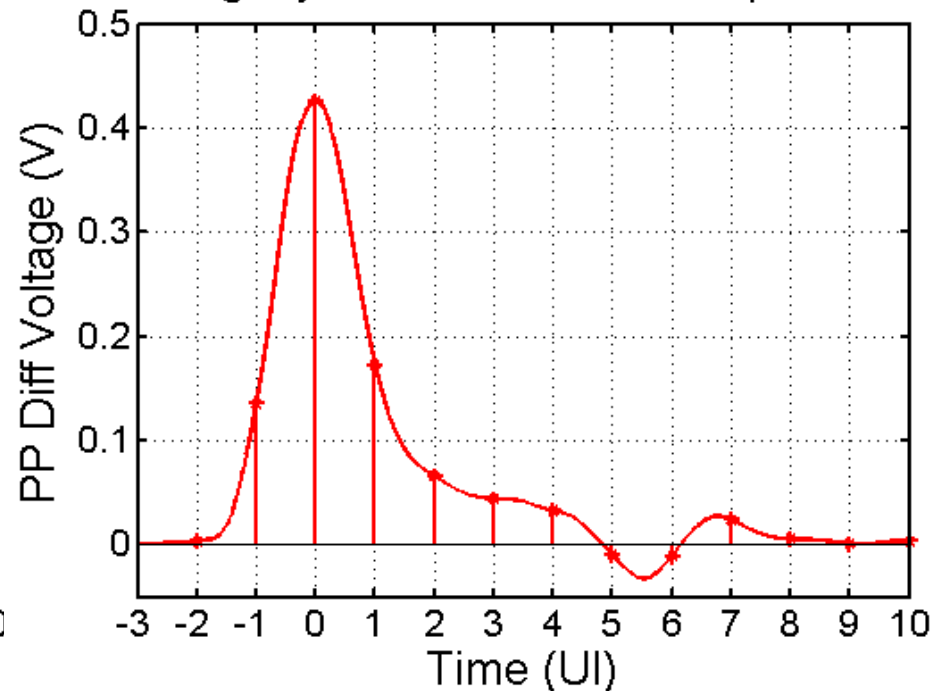
# Inter-Symbol Interference (ISI)

- Previous bits residual state can distort the current bit, resulting in inter-symbol interference (ISI)
- ISI is caused by
  - Reflections, Channel resonances, Channel loss (dispersion)

Legacy BP 5Gb/s Pulse Response

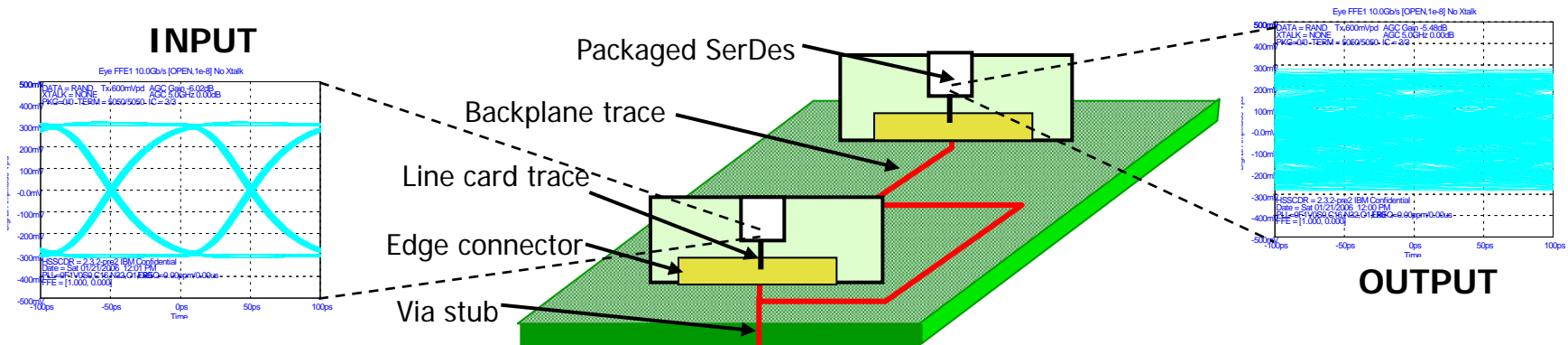


Legacy BP 5Gb/s Pulse Response



# ISI Impact

- At channel input (TX output), eye diagram is wide open
- As data pulses propagate through channel, they experience dispersion and have significant ISI
  - Result is a closed eye at channel output (RX input)



[Meghelli (IBM) ISSCC 2006]

# Next Time

---

- Channel pulse response model
- Modulation schemes